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A magnetoelectric screw dislocation interacting with a circular layered inclusion

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Abstract

Within the framework of the linear theory of magnetoelectroelasticity, the problem of a circular layered inclusion interacting with a generalized screw dislocation under remote anti-plane shear stress and in-plane magnetoelectric loads is investigated in this paper. The generalized dislocation can be located either in the matrix or in the circular layered inclusion. The layers are coaxial cylinders of annular cross-sections with arbitrary radii and different material properties. Using complex variable theory and the alternating technique, the solution of the present problem is expressed in terms of the solution of the corresponding homogeneous medium problem subjected to the same loading. Some numerical results are provided to investigate the influence of material combinations on the shear stress, electric field, magnetic and image force. These solutions can be used as Green's functions for the analysis of the corresponding magnetoelectric crack problem.

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1. Introduction

Due to magnetoelectroelastic coupling behavior, piezoelectric/piezomagnetic composite materials are extensively used to design actuators, sensors and other electronic products in modern technology. Combining two or more distinct constituents, composite materials can take advantage of each constituent and consequently exhibit a superior magnetoelectroelastic effect. Van Suchtelen (1972) first reported that piezoelectric/piezomagnetic composites may exhibit a new property viz. the magnetoelectroelastic coupling effect. Later, van Run et al. (1974) observed that the magnetoelectric coefficients of a BaTiO₃—CoFe₂O₄ composite are two orders higher than the highest known magnetoelectric coefficients of a Cr₂O₃ medium. Studies of the properties of piezoelectric/piezomagnetic composites have been carried out by numerous investigators in recent years. Nan (1994) and Huang and Kuo (1997) proposed micromechanics models to estimate the effective properties of fibrous piezoelectric/piezomagnetic composite materials. Li (2000) and Wu and Huang (2000) studied the inclusion and inhomogeneity problems and predicted the magnetoelectric coupling coefficients of the piezoelectric/piezomagnetic composites. Pan (2001) and Chen et al. (2002) proposed micromechanics models to evaluate the effective properties of plane, layered piezoelectric/piezomagnetic composite

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materials. Wang and Shen (2003) obtained analytical solutions for the problem of inclusions of arbitrary shape in magnetoelectroelastic composites.

Dislocation solutions or Green's functions can serve as kernel functions for use in the singular integral equation method. They are one of the well-established tools in the solution of numerous problems in the mechanics and physics of solids. For example, Li and Dunn (1998a; 1998b) used three-dimensional magnetoelectroelastic Green's functions to study inclusion problems. Green's functions for magnetoelectroelastic solids have been studied by several authors. Liu et al. (2001) gave Green's functions for an infinite two-dimensional anisotropic magnetoelectroelastic medium containing an elliptical cavity or a crack. Li (2002) obtained explicit expressions for the magnetoelectric Green's functions for a transversely isotropic medium and used them to analyze the magnetoelectric inclusion and inhomogeneity problems. Recently, Fang et al. (2005) discussed the interaction between a generalized screw dislocation with circular-arc interfacial rigid lines in magnetoelectroelastic solids. Hao and Liu (2006) investigated the interaction between a screw dislocation and a semi-infinite crack in a transversely isotropic magnetoelectroelastic bimaterial. Zheng et al. (2007) discussed the interaction between a generalized screw dislocation with circular-arc interfacial cracks along a circular inhomogeneity in magnetoelectroelastic solids.

Because of the presence of materials such as fiber coatings or transitional layers between the inclusion and the matrix, it is more reasonable to regard an interface as an interphase layer with finite thickness. The three-phase models provide accurate predictions of the effective moduli of composite materials, and can be used to study the influence of the interphase layer on the shear stress, electric field and magnetic. The importance of three-phase models in composite mechanics research has been described by Xiao and Chen (2000), Jiang and Cheung (2001), Sudak (2003) and Liu et al. (2003). In this paper, we study the magnetoelectroelastic interaction between a generalized screw dislocation and a circular layered inclusion. The proposed method is based upon the technique of analytic continuation. It is alternately applied across the two concentric interfaces in order to derive the trimaterial solution in a series form, from the corresponding homogeneous medium solution. Following the introduction, a complex representation of anti-plane magnetoelectroelasticity is provided in Section 2 and solutions to the corresponding homogeneous medium, circular inclusion and circular layered inclusion problems are provided in Sections 3, 4 and 5, respectively. Several numerical examples are given and discussed in detail in Section 6.

2. A complex representation of anti-plane magnetoelectroelasticity

Consider a magnetoelectroelastic composite composed of three dissimilar materials bonded along two concentric circular interfaces. Each component is assumed to have the same material orientation with x_3 being the poling direction (Fig. 1). In a class of magnetoelectroelastic materials capable of undergoing out-of-plane displacement w, the in-plane electric potential is φ and the in-plane magnetic potential is φ , the governing field equations can be simplified to

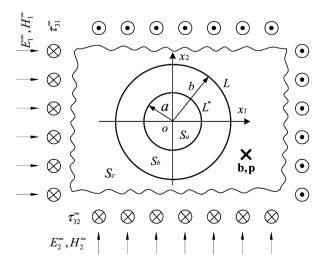


Fig. 1. A circular layered composite under magnetoelectromechanical loading.

$$c_{44}\nabla^2 w + e_{15}\nabla^2 \varphi + q_{15}\nabla^2 \phi = 0, (1)$$

$$e_{15}\nabla^2 w - \varepsilon_{11}\nabla^2 \varphi + d_{11}\nabla^2 \phi = 0, (2)$$

$$q_{15}\nabla^2 w - d_{11}\nabla^2 \varphi + \mu_{11}\nabla^2 \phi = 0, (3)$$

where ∇^2 is the two-dimensional Laplace operator, c_{44} is a elastic modulus, q_{15} a piezomagnetic constant, e_{15} a piezoelectric constant, ε_{11} a dielectric constant and d_{11} a magnetoelectric constant. According to Zheng et al. (2007), the shear stress components τ_{31} and τ_{32} , shear strain components γ_{31} and γ_{32} , electric displacement components D_1 and D_2 , electrical field components E_1 and E_2 , magnetic induction components E_1 and E_2 , and magnetic field components E_1 and E_2 and E_3 and E_4 and E_5 and E_7 and E_8 are E_8 and E_8 and E_8 and E_8 are E_8 and E_8 and E_8 and E_8 are E_8 and E_8 are E_8 and E_8 are E_8 are E_8 and E_8 are E_8 and E_8 are E_8 and E_8 are E_8 and E_8 are E_8 are E_8 are E_8 are E_8 and E_8 are E_8 are E_8 are E_8 and E_8 are E_8 are E_8 are E_8 are E_8 are E_8 are E_8 and E_8 are E_8 are E_8 are E_8 and E_8 are E_8

$$\begin{cases}
\gamma_{31} - i\gamma_{32} \\
-E_1 + iE_2 \\
-H_1 + iH_2
\end{cases} = \mathbf{f}'(z), \tag{6}$$

where Re denotes the real part and prime indicates differentiation with respect to the complex variable $z = x_1 + ix_2$. In order to express the boundary condition in terms of $\mathbf{f}(z)$ rather than its derivative $\mathbf{f}'(z)$, we integrate the traction t, the normal electric displacement D_n and the normal magnetic induction B_n as

$$\int \left\{ \begin{array}{l} t \\ D_n \\ B_n \end{array} \right\} ds = \operatorname{Im} \left[\mathbf{Cf}(z) \right], \tag{7}$$

where Im denotes the imaginary part of a complex function.

3. Solution of a homogeneous medium problem

Now we examine the homogeneous medium solution $\mathbf{f}_0(z)$ of a generalized screw dislocation in a homogeneous medium which is subjected to remote magnetoelectromechanical loadings. Let the generalized screw dislocation with vector $\mathbf{b} = [b_z, b_{\varphi}, b_{\phi}]^t$ and generalized line force $\mathbf{p} = [-p_z, p_{\varphi}, p_{\phi}]^t$ be applied at a point z_0 , where $b_z, b_{\varphi}, b_{\phi}$ represent the jump in displacement, electrical potential and magnetic potential across the slip plane respectively, and $p_z, p_{\varphi}, p_{\phi}$, represent a line force, a line electric charge and a line magnetic charge respectively. Referring to the work of Zheng et al. (2007), the complex potential corresponding to this homogeneous medium problem can be expressed as

$$\mathbf{f}_{0}(z) = \frac{1}{2\pi i} (\mathbf{b} + i\mathbf{C}^{-1}\mathbf{p}) \log(z - z_{0}) + \mathbf{C}^{-1}\mathbf{\Gamma}z$$

$$= \mathbf{D}\log(z - z_{0}) + \mathbf{C}^{-1}\mathbf{\Gamma}z,$$
(8)

where $\Gamma = [\tau_{31}^{\infty} - i\tau_{32}^{\infty}, D_1^{\infty} - iD_2^{\infty}, B_1^{\infty} - iB_2^{\infty}]^t$ represents the remote magnetoelectromechanical loading.

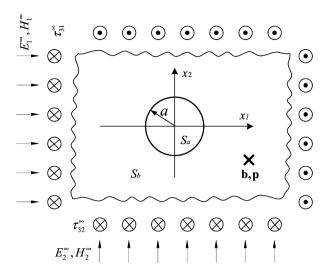


Fig. 2. A circular inclusion under magnetoelectromechanical loading.

4. A circular inclusion in a magnetoelectroelastic solid

Consider a two-phase magnetoelectroelastic composite bonded along the circular interface r = a with a magnetoelectric loading applied in the matrix (Fig. 2). Our objective is to construct a bi-material solution in the form

$$\mathbf{f}(z) = \begin{cases} \mathbf{f}_a(z), & z \in S_a, \\ \mathbf{f}_b(z) + \mathbf{f}_0(z), & z \in S_b, \end{cases}$$
(9)

where S_a , the inner region, and S_b , the outer region, are occupied by material a and b respectively, and $\mathbf{f}_0(z)$ represents the solution corresponding to the homogeneous medium which is holomorphic in the entire domain except for some singular points. In order to express $\mathbf{f}_a(z)$ and $\mathbf{f}_b(z)$ (holomorphic in S_a and S_b , respectively) in terms of $\mathbf{f}_0(z)$, continuity of displacement, electric potential and magnetic potential across the interface, by the analytical continuation theorem, is used to yield the requirement

$$\begin{cases}
\mathbf{f}_{a}(z) - \overline{\mathbf{f}}_{b}\left(\frac{a^{2}}{z}\right) - \mathbf{f}_{0}(z) = 0, & z \in S_{a}, \\
\overline{\mathbf{f}}_{a}\left(\frac{a^{2}}{z}\right) - \mathbf{f}_{b}(z) - \overline{\mathbf{f}}_{0}\left(\frac{a^{2}}{z}\right) = 0, & z \in S_{b}.
\end{cases}$$
(10)

Continuity of traction and normal component of the electric displacement and magnetic induction, by the same arguments, results in

$$\begin{cases}
\mathbf{C}_{a}\mathbf{f}_{a}(z) + \mathbf{C}_{b}\overline{\mathbf{f}_{b}}\left(\frac{a^{2}}{z}\right) - \mathbf{C}_{b}\mathbf{f}_{0}(z) = 0, & z \in S_{a}, \\
\mathbf{C}_{a}\overline{\mathbf{f}_{a}}\left(\frac{a^{2}}{z}\right) + \mathbf{C}_{b}\mathbf{f}_{b}(z) - \mathbf{C}_{b}\overline{\mathbf{f}_{0}}\left(\frac{a^{2}}{z}\right) = 0, & z \in S_{b},
\end{cases}$$
(11)

where

$$\mathbf{C}_{a} = \begin{bmatrix} c_{44}^{(a)} & e_{15}^{(a)} & q_{15}^{(a)} \\ e_{15}^{(a)} & -\varepsilon_{11}^{(a)} & -d_{11}^{(a)} \\ q_{15}^{(a)} & -d_{11}^{(a)} & -\mu_{11}^{(a)} \end{bmatrix},$$

$$\mathbf{C}_b = \begin{bmatrix} c_{44}^{(b)} & e_{15}^{(b)} & q_{15}^{(b)} \\ e_{15}^{(b)} & -\varepsilon_{11}^{(b)} & -d_{11}^{(b)} \\ q_{15}^{(b)} & -d_{11}^{(b)} & -\mu_{11}^{(b)} \end{bmatrix}.$$

From Eqs. (10) and (11), we have

$$\begin{cases}
\mathbf{f}_{a}(z) = \boldsymbol{\alpha}_{ab}\mathbf{f}_{0}(z), \\
\mathbf{f}_{b}(z) = \boldsymbol{\beta}_{ab}\overline{\mathbf{f}_{0}}\left(\frac{a^{2}}{z}\right),
\end{cases}$$
(12)

where

$$\boldsymbol{\alpha}_{ab} = 2(\mathbf{C}_a + \mathbf{C}_b)^{-1}\mathbf{C}_b,\tag{13}$$

$$\boldsymbol{\beta}_{ab} = (\mathbf{C}_a + \mathbf{C}_b)^{-1} (\mathbf{C}_b - \mathbf{C}_a). \tag{14}$$

For a magnetoelectric loading applied to the inner region, since the homogeneous medium solution is holomorphic in neither the region S_a nor the region S_b , it should be rewritten as

$$\mathbf{f}_0(z) = \mathbf{D}_a \log z + \mathbf{f}_0^*(z),$$

where \mathbf{D}_a is defined as in Eq. (8), but the magnetoelectroelastic constants involved in \mathbf{D}_a are for material a, and

$$\mathbf{f}_0^*(z) = \mathbf{D}_a \log \left(1 - \frac{z_0}{z} \right),$$

which is holomorphic in the outer region S_b . By applying the continuity conditions of $Re[\mathbf{f}(z)]$ and $Im[\mathbf{Cf}(z)]$ along the interface, and using the method of analytic continuation, one can obtain

$$\mathbf{f}(z) = \begin{cases} \mathbf{f}_0(z) + \mathbf{f}_a(z), & z \in S_a, \\ \mathbf{D}_b \log\left(\frac{z}{a}\right) + \mathbf{D}_a \log(a) + \mathbf{f}_b(z), & z \in S_b, \end{cases}$$
(15)

where

$$\begin{cases}
\mathbf{f}_{a}(z) = \boldsymbol{\beta}_{ba} \overline{\mathbf{f}_{0}^{*}} \left(\frac{a^{2}}{z}\right), \\
\mathbf{f}_{b}(z) = \boldsymbol{\alpha}_{ba} \mathbf{f}_{0}^{*}(z),
\end{cases} (16)$$

and \mathbf{D}_b is defined as in Eq. (8), but the magnetoelectroelastic constants involved in \mathbf{D}_b are for material b.

5. A circular layered inclusion in a magnetoelectroelastic solid

In this section, we consider a three-phase magnetoelectroelastic composite (Fig. 1), whose cross-section consists of a circular ring S_b , with outer and inner radii designated by b and a, respectively. Its inner and outer boundaries are perfectly bonded to a circular inclusion S_a and a matrix S_c , respectively.

5.1. Case I: A generalized screw dislocation located in the interphase layer S_b

A series solution for the present three-phase magnetoelectroelastic composite with a generalized screw dislocation located in the interphase layer S_b is assumed, in the form

$$\mathbf{f}(z) = \begin{cases} \sum_{n=1}^{\infty} \mathbf{f}_{an}(z), & z \in S_a, \\ \mathbf{f}_0(z) + \sum_{n=1}^{\infty} \mathbf{f}_n(z) + \sum_{n=1}^{\infty} \mathbf{f}_{bn}(z), & z \in S_b, \\ \mathbf{D}_c \log\left(\frac{z}{b}\right) + \mathbf{D}_b \log(b) + \mathbf{f}_{c0}(z) + \sum_{n=1}^{\infty} \mathbf{f}_{cn}(z), & z \in S_c, \end{cases}$$
(17)

where \mathbf{D}_c is defined as \mathbf{D}_a in Section 4, but the magnetoelectroelastic constants involved in \mathbf{D}_c are for material c. Now, it is required to solve for $\mathbf{f}_{c0}(z)$, $\mathbf{f}_{an}(z)$, $\mathbf{f}_{bn}(z)$, $\mathbf{f}_{cn}(z)$ and $\mathbf{f}_n(z)$ (n = 1, 2, 3, ...), analytic in their respective regions, in terms of $\mathbf{f}_0(z)$, by the procedure as follows.

Step 1. Analytic continuation across the interface L.

First, we regard regions S_a and S_b as being composed of the same material b, and region S_c of material c. Since $\mathbf{f}_0(z)$ is a homogeneous medium solution and $\mathbf{f}_{c0}(z)$ and $\mathbf{f}_1(z)$ are introduced to satisfy the continuity conditions across the interface L, Eqs. (15) and (16) lead to

$$\begin{cases}
\mathbf{f}_1(z) = \boldsymbol{\beta}_{cb} \overline{\mathbf{f}_0^*} \left(\frac{b^2}{z} \right), & z \in S_a \cup S_b, \\
\mathbf{f}_{c0}(z) = \boldsymbol{\alpha}_{cb} \mathbf{f}_0^*(z), & z \in S_c,
\end{cases}$$
(18)

where $\mathbf{f}_0^*(z) = \mathbf{D}_b \log(1 - z_0/z)$. Since we assume that region S_a is made up of material b, the fields produced by $\mathbf{f}_1(z)$ cannot satisfy the continuity conditions at the interface L^* , which lies between material a and b.

Step 2. Analytic continuation across the interface L^* .

Next, we assume that region S_a is composed of material a and regions S_b and S_c are regarded as made up of the same material b. $\mathbf{f}_1(z)$ in Eq. (18) having the singular points in $S_b \cup S_c$ is treated as a homogeneous solution of material b. As in Eqs. (9) and (12), $\mathbf{f}_{a1}(z)$ and $\mathbf{f}_{b1}(z)$ are introduced to satisfy the continuity conditions across the interface L^* as

$$\begin{cases}
\mathbf{f}_{a1}(z) = \boldsymbol{\alpha}_{ab} \left[\mathbf{f}_{1}(z) + \mathbf{f}_{0}(z) \right], & z \in S_{a}, \\
\mathbf{f}_{b1}(z) = \boldsymbol{\beta}_{ab} \left[\overline{\mathbf{f}_{1}} \left(\frac{a^{2}}{z} \right) + \overline{\mathbf{f}_{0}} \left(\frac{a^{2}}{z} \right) \right], & z \in S_{b} \cup S_{c}.
\end{cases}$$
(19)

Since this result is based on the assumption that region S_c is made up of material b, the field produced by $\mathbf{f}_{b1}(z)$ cannot satisfy the continuity conditions at the interface L.

Step 3. Analytic continuation across the interface L.

We again assume regions S_a and S_b be made up of the same material b and region S_c of material c. As in Eqs. (15) and (16), $\mathbf{f}_{b1}(z)$ is taken to be a homogeneous solution of material b, and $\mathbf{f}_{c1}(z)$ and $\mathbf{f}_{c1}(z)$ are introduced to satisfy the continuity conditions across the interface L. Accordingly, it can be shown

$$\begin{cases}
\mathbf{f}_{2}(z) = \boldsymbol{\beta}_{cb} \overline{\mathbf{f}_{b1}} \left(\frac{b^{2}}{z} \right), & z \in S_{a} \cup S_{b}, \\
\mathbf{f}_{c1}(z) = \boldsymbol{\alpha}_{cb} \mathbf{f}_{b1}(z), & z \in S_{c}.
\end{cases}$$
(20)

Similarly, the fields produced by $\mathbf{f}_2(z)$ do not satisfy the conditions at the interface L^* .

Step 4. Repetitions of steps 2 and 3.

By repeating steps 2 and 3 for n = 2, 3, 4, ..., one can express all the functions $\mathbf{f}_{c0}(z)$, $\mathbf{f}_{an}(z)$, $\mathbf{f}_{bn}(z)$, and $\mathbf{f}_{cn}(z)$ (n = 1, 2, 3, ...) in terms of $\mathbf{f}_0(z)$ as

$$\mathbf{f}(z) = \begin{cases} \boldsymbol{\alpha}_{ab} \sum_{n=0}^{\infty} (\boldsymbol{\beta}_{cb} \boldsymbol{\beta}_{ab})^{n} \left[\mathbf{f}_{0} \left(\frac{a^{2n}}{b^{2n}} z \right) + \boldsymbol{\beta}_{cb} \overline{\mathbf{f}}_{0}^{*} \left(\frac{b^{2n+2}}{a^{2n}} \frac{1}{z} \right) \right], & z \in S_{a}, \\ \sum_{n=0}^{\infty} (\boldsymbol{\beta}_{cb} \boldsymbol{\beta}_{ab})^{n} \left[\mathbf{f}_{0} \left(\frac{a^{2n}}{b^{2n}} z \right) + \boldsymbol{\beta}_{cb} \overline{\mathbf{f}}_{0}^{*} \left(\frac{b^{2n+2}}{a^{2n}} \frac{1}{z} \right) \right] \\ + \boldsymbol{\beta}_{ab} \sum_{n=0}^{\infty} (\boldsymbol{\beta}_{cb} \boldsymbol{\beta}_{ab})^{n} \left[\overline{\mathbf{f}}_{0} \left(\frac{a^{2n+2}}{b^{2n}} \frac{1}{z} \right) + \boldsymbol{\beta}_{cb} \mathbf{f}_{0}^{*} \left(\frac{b^{2n+2}}{a^{2n+2}} z \right) \right], & z \in S_{b}, \end{cases}$$

$$\mathbf{D}_{c} \log \left(\frac{z}{b} \right) + \mathbf{D}_{b} \log(b) + \boldsymbol{\alpha}_{cb} \mathbf{f}_{0}^{*} (z) \\ + \boldsymbol{\alpha}_{cb} \boldsymbol{\beta}_{ab} \sum_{n=0}^{\infty} (\boldsymbol{\beta}_{cb} \boldsymbol{\beta}_{ab})^{n} \left[\overline{\mathbf{f}}_{0} \left(\frac{a^{2n+2}}{b^{2n}} \frac{1}{z} \right) + \boldsymbol{\beta}_{cb} \mathbf{f}_{0}^{*} \left(\frac{b^{2n+2}}{a^{2n+2}} z \right) \right], & z \in S_{c}. \end{cases}$$

$$(21)$$

5.2. Case II: A generalized screw dislocation located in the matrix S_c

Using the same procedure as case I, the case in which the generalized screw dislocation located in the matrix S_c has the following solution

$$\mathbf{f}(z) = \begin{cases} \boldsymbol{\alpha}_{ab} \sum_{n=0}^{\infty} (\boldsymbol{\beta}_{cb} \boldsymbol{\beta}_{ab})^n \boldsymbol{\alpha}_{bc} \mathbf{f}_0 \left(\frac{a^{2n}}{b^{2n}} z \right), & z \in S_a, \\ \sum_{n=0}^{\infty} (\boldsymbol{\beta}_{cb} \boldsymbol{\beta}_{ab})^n \boldsymbol{\alpha}_{bc} \mathbf{f}_0 \left(\frac{a^{2n}}{b^{2n}} z \right) + \boldsymbol{\beta}_{ab} \sum_{n=0}^{\infty} (\boldsymbol{\beta}_{cb} \boldsymbol{\beta}_{ab})^n \boldsymbol{\alpha}_{bc} \overline{\mathbf{f}_0} \left(\frac{a^{2n+2}}{b^{2n}} \frac{1}{z} \right), & z \in S_b, \\ \mathbf{f}_0(z) + \boldsymbol{\beta}_{bc} \overline{\mathbf{f}_0} \left(\frac{b^2}{z} \right) + \boldsymbol{\alpha}_{cb} \boldsymbol{\beta}_{ab} \sum_{n=0}^{\infty} (\boldsymbol{\beta}_{cb} \boldsymbol{\beta}_{ab})^n \boldsymbol{\alpha}_{bc} \overline{\mathbf{f}_0} \left(\frac{a^{2n+2}}{b^{2n}} \frac{1}{z} \right), & z \in S_c. \end{cases}$$

5.3. Case III: A generalized screw dislocation located in the inner inclusion S_a

Similarly, the solution for the case in which the generalized screw dislocation located in the inner inclusion S_a can be derived as

$$\mathbf{f}(z) = \begin{cases} \mathbf{f}_{0}(z) + \boldsymbol{\beta}_{ba} \overline{\mathbf{f}_{0}^{*}} \left(\frac{a^{2}}{z}\right) + \boldsymbol{\alpha}_{ab} \boldsymbol{\beta}_{cb} \sum_{n=0}^{\infty} (\boldsymbol{\beta}_{ab} \boldsymbol{\beta}_{cb})^{n} \boldsymbol{\alpha}_{bc} \overline{\mathbf{f}_{0}^{*}} \left(\frac{b^{2n+2}}{a^{2n}} \frac{1}{z}\right), & z \in S_{a}, \\ \mathbf{D}_{b} \log \left(\frac{z}{a}\right) + \mathbf{D}_{a} \log a + \sum_{n=0}^{\infty} (\boldsymbol{\beta}_{ab} \boldsymbol{\beta}_{cb})^{n} \boldsymbol{\alpha}_{ba} \mathbf{f}_{0}^{*} \left(\frac{b^{2n}}{a^{2n}} z\right) \\ + \boldsymbol{\beta}_{cb} \sum_{n=0}^{\infty} (\boldsymbol{\beta}_{ab} \boldsymbol{\beta}_{cb})^{n} \boldsymbol{\alpha}_{bc} \overline{\mathbf{f}_{0}^{*}} \left(\frac{b^{2n+2}}{a^{2n}} \frac{1}{z}\right), & z \in S_{b}, \\ \mathbf{D}_{c} \log \left(\frac{z}{b}\right) + \mathbf{D}_{b} \log \left(\frac{b}{a}\right) + \mathbf{D}_{a} \log a + \boldsymbol{\alpha}_{cb} \sum_{n=0}^{\infty} (\boldsymbol{\beta}_{ab} \boldsymbol{\beta}_{cb})^{n} \boldsymbol{\alpha}_{ba} \mathbf{f}_{0}^{*} \left(\frac{b^{2n}}{a^{2n}} z\right), & z \in S_{c}. \end{cases}$$

$$(23)$$

Eqs. (21), (22) and (23) give a general series solution of the problem associated with the three-phase circular layered composite under arbitrary loading as the corresponding homogeneous medium solution $\mathbf{f}_0(z)$ is solved. Applying the derived analytical solutions and assuming the dislocation density along the crack as an unknown function, one can obtain the standard singular integral equations for the corresponding magnetoelectric crack problem. A numerical solution of the dislocation density function can be obtained by solving the resulting linear algebraic equations. Similar numerical processes can be found in the literature, e.g. Chen and Hesebe (1992), Chao and Shen (1995) and Wang and Zhong (2003).

6. Results and discussion

In this section, the fundamental series solution derived in the preceding section will be used to analyze the following examples associated with a circular layered piezoelectric/piezomagnetic composite subjected to magnetoelectromechanical loading. Numerical results are obtained for the composite BaTiO₃–CoFe₂O₄, the constituents of which are: piezoelectric BaTiO₃ and piezomagnetic CoFe₂O₄, whose material constants are given as Zheng et al. (2007)

- piezoelectric BaTiO₃: $c_{44} = 43 \times 10^9 \text{ N/m}^2$, $e_{15} = 11.6 \text{ C/m}^2$, $q_{15} = 0$, $\varepsilon_{11} = 11.2 \times 10^{-9} \text{ C}^2/\text{ N m}^2$, $d_{11} = 0$, $\mu_{11} = 5.0 \times 10^{-6} \text{ N s}^2/\text{C}^2$.
- piezomagnetic CoFe₂O₄: $c_{44} = 45.3 \times 10^9 \text{ N/m}^2$, $e_{15} = 0$, $q_{15} = 550 \text{ N/A m}$, $\varepsilon_{11} = 0.08 \times 10^{-9} \text{ C}^2/\text{N m}^2$, $d_{11} = 0$, $\mu_{11} = -590 \times 10^{-6} \text{ N s}^2/\text{C}^2$.

6.1. Uniform far-field magnetoelectroelastic loading

We first consider the problem of a tri-magnetoelectric composite subjected to a uniform far-field anti-plane shear τ_{31}^{∞} . Figs. 3–5 show the variations of the shear stress τ_{31} , electric displacement D_1 and magnetic induction

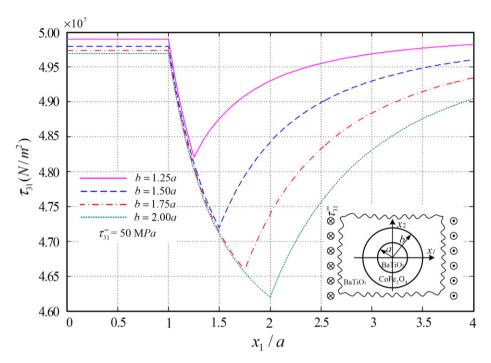


Fig. 3. The variation of the shear stress τ_{31} along the real axis x_1 .

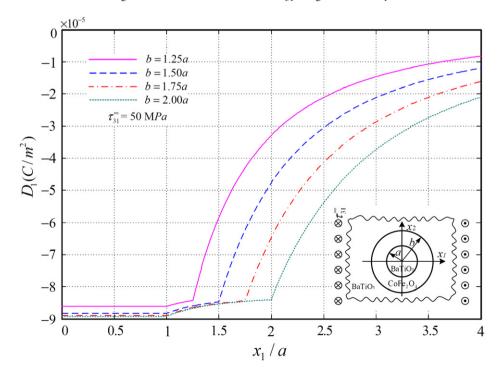


Fig. 4. The variation of the electric displacement D_1 along the real axis x_1 .

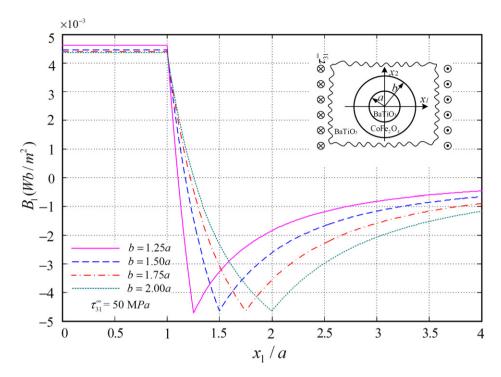


Fig. 5. The variation of the magnetic induction B_1 along the real axis x_1 .

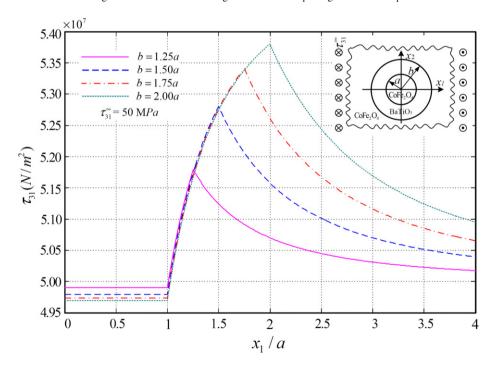


Fig. 6. The variation of the shear stress τ_{31} along the real axis x_1 .

 B_1 along the real axis x_1 for the case when the inner inclusion and the matrix are piezoelectric BaTiO₃ and the interphase layer is piezomagnetic CoFe₂O₄. Figs. 6–8 show the variations of the shear τ_{31} , electric displacement D_1 and magnetic induction B_1 along the real axis x_1 for the case when the inner inclusion and the matrix are piezomagnetic CoFe₂O₄ and the interphase layer is piezoelectric BaTiO₃. From these figures, we make the following deductions:

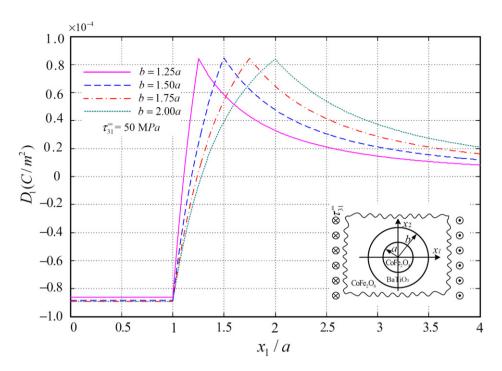


Fig. 7. The variation of the electric displacement D_1 along the real axis x_1 .

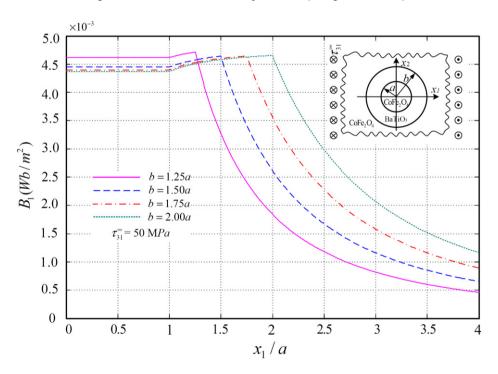


Fig. 8. The variation of the magnetic induction B_1 along the real axis x_1 .

- 1. The piezoelectric/piezomagnetic composite subjected to a far-field mechanical loading τ_{31}^{∞} will exhibit the magnetoelectroelastic coupling effect.
- 2. The shear stress τ_{31} , electric displacement D_1 and magnetic induction B_1 produced in the inner inclusion are uniform.

- 3. The shear stress τ_{31} , electric displacement D_1 and magnetic induction B_1 are continuous across the interface, which conforms to the boundary conditions.
- 4. When the interphase layer is piezomagnetic $CoFe_2O_4$, the variation of magnetic induction B_1 in the interphase layer is higher than when the interphase layer is piezoelectric BaTiO₃. Similarly, when the interphase layer is piezoelectric BaTiO₃, the variation of electric displacement D_1 in the interphase layer is higher than those when the interphase layer is piezomagnetic $CoFe_2O_4$.
- 5. For large x_1 , the shear stress τ_{31} , electric displacement D_1 and magnetic induction B_1 will tend to the applied loading. Further, $\tau_{31} = \tau_{31}^{\infty}$, $D_1 = 0$ and $B_1 = 0$, which also conforms to the boundary conditions.

6.2. Image force on the generalized screw dislocation

The image force acting on the dislocation is an important physical quantity for understanding the interaction of a dislocation with inhomogeneities. The image force can be calculated by means of the generalized Peach–Koehler formula, by Pak (1990).

$$F_1 = b_z \tau_{32}^T + b_{\varphi} D_2^T + b_{\phi} B_2^T,$$

$$F_2 = -b_z \tau_{31}^T - b_{\varphi} D_1^T - b_{\phi} B_1^T,$$
(24)

where τ_{31}^T , τ_{32}^T , D_1^T , D_2^T , B_1^T and B_2^T are the perturbation shear stress, electric displacement and magnetic induction components at the dislocation. Suppose a generalized screw dislocation with Burgers vector b_z or electric potential jump b_{φ} is located at the point z_0 along the real axis in the matrix or in the circular layered inclusion. Define the normalized force on the dislocation as $F_{10}(b_z) = 2\pi a F_1/c_{44}b_z^2$, $F_{20}(b_z) = 2\pi a F_2/c_{44}b_z^2$, $F_{10}(b_{\varphi}) = 2\pi a F_1/c_{11}b_{\varphi}^2$, $F_{20}(b_{\varphi}) = 2\pi a F_2/c_{11}b_{\varphi}^2$. Figs. 9 and 10 respectively show the variation of normalized image force $F_{10}(b_z)$ with location of the dislocation, for the case (a) when the inner inclusion and the matrix are piezoelectric (BaTiO₃) and the interphase layer is piezomagnetic (CoFe₂O₄) and (b) when the inner inclusion and the matrix are piezomagnetic (CoFe₂O₄) and the interphase layer is piezoelectric (BaTiO₃). It is seen that the interfaces will repel the dislocation having Burgers vector, b_z , located in any region. The origin, a point near the middle of the interphase layer and the point at infinity are three stable equilibrium positions in which the image forces are equal to zero. Figs. 11

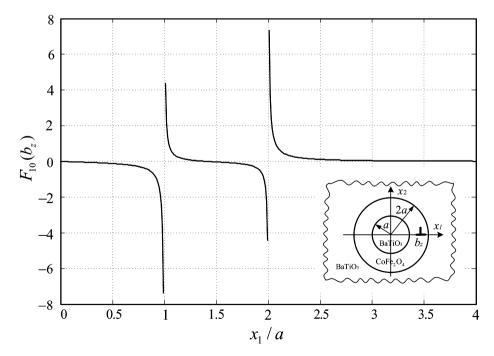


Fig. 9. The variation of normalized image force $F_{10}(b_z)$ with location of the dislocation.

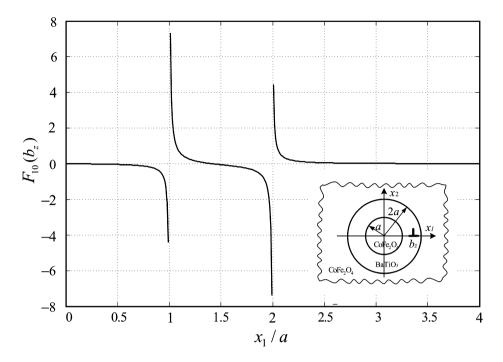


Fig. 10. The variation of normalized image force $F_{10}(b_z)$ with location of the dislocation.

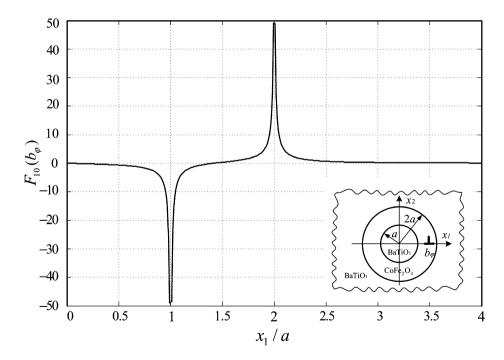


Fig. 11. The variation of normalized image force $F_{10}(b_{\varphi})$ with location of the dislocation.

and 12 show the variations of normalized image force $F_{10}(b_{\varphi})$ with respect to the location of the dislocation. It is seen that the interfaces will attract the dislocation having Burgers vector, b_{φ} , located in the interphase layer. A point near the middle of the interphase layer is an unstable equilibrium position when the interphase layer is piezomagnetic

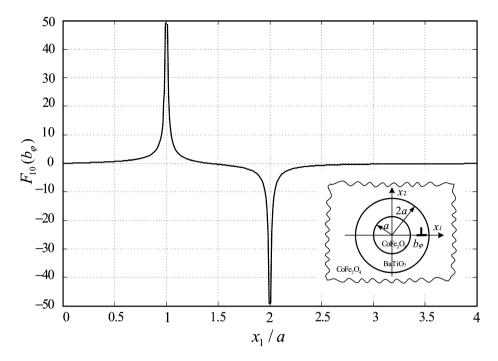


Fig. 12. The variation of normalized image force $F_{10}(b_{\varphi})$ with location of the dislocation.

CoFe₂O₄, while the interfaces will repel the dislocation, b_{φ} , located in the interphase layer. A point near the middle of the interphase layer is a stable equilibrium position when the interphase layer is piezoelectric BaTiO₃.

7. Concluding remarks

In this paper, the problem of a circular layered composite under magnetoelectromechanical loading is solved. Based on the method of analytic continuation and the alternating technique, the solution is obtained as a transformation of the solution to the corresponding homogeneous medium solution. Numerical results show that for the special case of uniform remote loading, the shear stress, electric displacement and magnetic induction produced in the inner inclusion are uniform. When the interphase layer is piezomagnetic $CoFe_2O_4$ the variations of magnetic induction B_1 in the interphase layer are higher than those when the interphase layer is piezoelectric $BaTiO_3$. Similarly, when the interphase layer is piezoelectric $BaTiO_3$ the variations of electric displacement D_1 in the interphase layer are higher than those when the interphase layer is piezomagnetic $CoFe_2O_4$. These important phenomena show the feasibility of building a very sensitive sensor in magnetoelectric composites.

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